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# Fundamentals of Electromagnetic Fields and Waves: I

Fall 2007, EE 30348, Electrical Engineering, University of Notre Dame

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## Final Exam Solutions (30 Points)

- Formulae are given at the end of this questionnaire.
- Please show your steps clearly and sketch figures wherever necessary.
- Points will be awarded for correct steps shown in the solutions.
- There are **four** questions in this exam. Please attempt all questions.

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### Problem 1 (10 Points)

Answer the following short questions.

a) (**2 Points**) Can the following vector fields represent a *static* Electric field?

i)  $\mathbf{A} = 3\phi\mathbf{a}_\phi$ .

**Solution:** For a vector field to represent a static electric field, Faraday's law dictates that  $\nabla \times \mathbf{E} = 0$ .  $\nabla \times \mathbf{A} = \frac{1}{\rho} \frac{\partial}{\partial \rho} (3\phi\rho)\mathbf{a}_z = \frac{3\phi}{\rho}\mathbf{a}_z \neq 0$ . Therefore  $\mathbf{A}$  cannot represent a static electric field.

ii)  $\mathbf{B} = x\mathbf{a}_x + y\mathbf{a}_y$ .

**Solution:** By the same argument, since  $\nabla \times \mathbf{B} = 0$ ,  $\mathbf{B}$  can represent a static electric field.

b) (**3 Points**) The electrostatic potential distribution in the region ( $x \geq 0, y \geq 0, z \geq 0$ ) which is filled with a material of dielectric constant  $\epsilon$  is given by

$$V(x, y, z) = V_0 e^{-x/L_x} e^{-y/L_y} e^{-z/L_z}$$

in Volts. Find the point  $(x, y, z)$  where the volume charge density  $\rho_v(x, y, z)$  reaches the maximum magnitude, and the volume charge density  $|\rho_v|_{max}$  at that point.

**Solution:** By Poisson's equation, the volume charge density at any point is related to the electrostatic potential by the relation  $\rho_v = -\epsilon \nabla^2 V = -\epsilon \left( \frac{1}{L_x^2} + \frac{1}{L_y^2} + \frac{1}{L_z^2} \right) V_0 e^{-x/L_x} e^{-y/L_y} e^{-z/L_z}$ . Clearly, it reaches a maximum at  $(0, 0, 0)$ , and the charge density at that point is  $|\rho_v|_{max} = \epsilon \left( \frac{1}{L_x^2} + \frac{1}{L_y^2} + \frac{1}{L_z^2} \right) V_0$ .

c) (**2 Points**) The region between two parallel plates of a capacitor is filled with a material of dielectric constant  $\epsilon$  and conductivity  $\sigma$ . Find the maximum permissible value of the conductivity such that any charge put on the plates at time  $t = 0$  can be stored for at least  $t_1$  seconds before reducing to 1/100 of its initial value. As discussed in class, the charge decays exponentially, with the time constant given by the characteristic dielectric relaxation time  $\tau_d = \epsilon/\sigma$ .

**Solution:** Since the charge decays as  $Q(t) = Q(0)e^{-t/\tau_d}$ , it is clear that to meet the requirement, we must have  $Q(t_1) = Q(0)e^{-t_1/\tau_d} \geq Q(0)/100 \rightarrow \tau_d \geq t_1/\ln 100 \rightarrow \sigma \leq \frac{\epsilon}{t_1} \ln 100$ .

d) (**3 Points**) Show that the skin depth  $\delta$  for an electromagnetic wave is

i) independent of the frequency of the wave if it propagates in a region of very low conductivity<sup>1</sup>, &

ii) goes as the inverse square root of the frequency of the wave ( $\delta \propto 1/\sqrt{\omega}$ ) if it propagates in a region of very high conductivity.

**Solution:** Since  $\alpha = \omega \sqrt{\frac{\mu\epsilon}{2} [\sqrt{1 + (\frac{\sigma}{\omega\epsilon})^2} - 1]}^{1/2}$ , for  $(\sigma/\omega\epsilon) \ll 1$ ,  $\alpha \approx \omega \sqrt{\frac{\mu\epsilon}{2} [1 + \frac{1}{2}(\frac{\sigma}{\omega\epsilon})^2 - 1]}^{1/2} \approx \frac{1}{2} \sqrt{\frac{\mu\sigma^2}{\epsilon}}$ . Hence the skin depth for EM wave propagation in a low conductivity material is given by  $\delta = \frac{1}{\alpha} \approx \sqrt{\frac{4\epsilon}{\mu\sigma^2}}$ , which is clearly not dependent on the frequency  $\omega$  of the wave.

On the other hand, for a highly conductive medium,  $(\sigma/\omega\epsilon) \gg 1$ ,  $\alpha \approx \omega \sqrt{\frac{\mu\epsilon}{2} [\sqrt{(\frac{\sigma}{\omega\epsilon})^2}]^{1/2}} \approx \sqrt{\frac{\mu\sigma\omega}{2}}$ . Hence the skin depth for EM wave propagation in a low conductivity material is given by  $\delta = \frac{1}{\alpha} \approx \sqrt{\frac{2}{\mu\sigma\omega}}$ , which goes as  $1/\sqrt{\omega}$ .

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<sup>1</sup>you might need the approximation  $\sqrt{1+x} \approx 1+x/2$  for  $x \ll 1$ .

**Problem 2 (6 Points)**

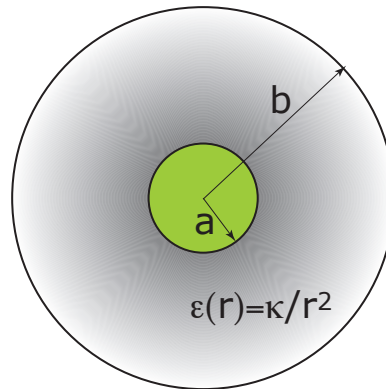


Figure 1: Problem (2)

A spherical capacitor has inner radius  $a$  and outer radius  $b$ . The region between the two spheres is filled with an *inhomogenous* dielectric medium, with dielectric constant given by  $\epsilon(r) = \kappa/r^2$  (see Figure 1). Find the capacitance of the structure. Show all steps.

**Solution:**

To find the capacitance, assume that we put  $+Q$  charge on the inner sphere, and  $-Q$  on the outer one. If we can find the voltage difference between the two conductors, the capacitance will be given by  $C = Q/V$ . The electric flux density  $\mathbf{D}$  can be found in the region  $a < r < b$  by applying the integral form of Gauss's law:  $\oint \mathbf{D} \cdot d\mathbf{S} = Q_{encl} \rightarrow \mathbf{D}(r) = \frac{Q}{4\pi r^2} \mathbf{a}_r$ . Then, the electric field is given by  $\mathbf{E}(r) = \frac{\mathbf{D}(r)}{\epsilon(r)} = \frac{Q}{4\pi\kappa} \mathbf{a}_r$ , using  $\epsilon(r) = \kappa/r^2$ . Then, the voltage difference between the two conductors is given by  $V_{ab} = \int_a^b \mathbf{E}(r) \cdot d\mathbf{l} = \frac{Q}{4\pi\kappa}(b - a)$ , and the capacitance is given by  $C = Q/V_{ab} = \frac{4\pi\kappa}{b-a}$ .

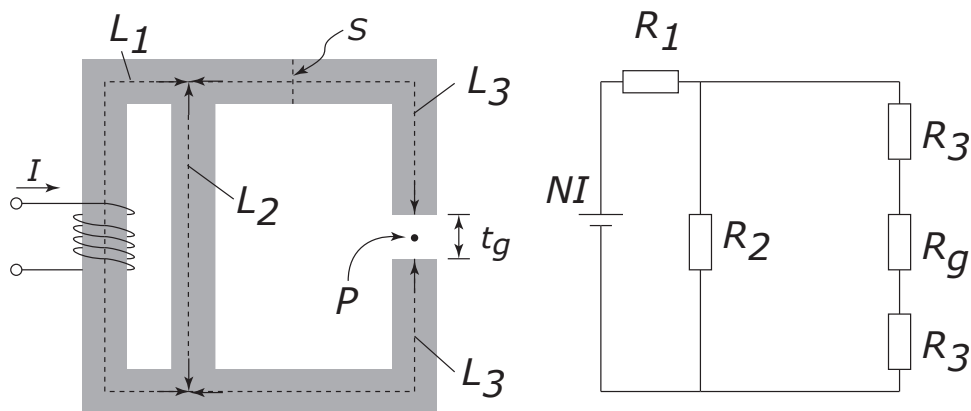


Figure 2: Problem (3)

### Problem 3 (6 Points)

Refer to the magnetic circuit shown in Figure 3. The ferromagnetic core has a permeability  $\mu$ , and the geometrical dimensions are shown (the cross-sectional area  $s$  is assumed to be uniform throughout the length of the core). You are given a current source which can deliver a current of  $I$  Amperes, and a long wire. Design the windings (i.e., how many turns  $N$  would you need) to create a magnetic field  $B_0$  at point  $P$  in the air gap? Find the answer in terms of the variables given. Neglect any flux leakage from the magnetic core and fringing fields in the air gap.

#### Solution:

The equivalent electrical circuit is shown in the figure above. The following are the corresponding reluctances:  $R_1 = L_1/\mu S$ ,  $R_2 = L_2/\mu S$ ,  $R_3 = L_3/\mu S$ , and  $R_g = t_g/\mu_0 S$ . The magnetomotive force is  $NI$ , which will create a net magnetic flux in the branch with the air gap equal to

$$\psi_m^{gap} = B_0 S = \frac{R_2}{R_2 + 2R_3 + R_g} \cdot \frac{NI}{R_1 + R_2 || (R_g + 2R_3)}. \quad (1)$$

Here  $a || b = (\frac{1}{a} + \frac{1}{b})^{-1}$ , the parallel addition of resistors. Clearly, to produce a magnetic field  $B_0$  in the air gap, the number of turns necessary is given by

$$N = \frac{R_2 + 2R_3 + R_g}{R_2} \cdot [R_1 + R_2 || (R_g + 2R_3)] \cdot \frac{B_0 S}{I}. \quad (2)$$

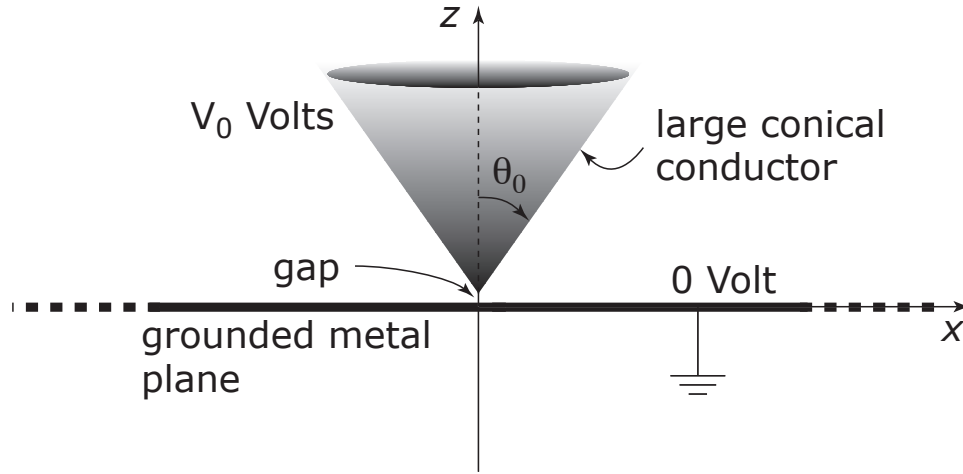


Figure 3: Problem (4)

**Problem 4 (8 Points)**

Refer to Figure 3. A large conducting cone (angle  $\theta_0$ ) is placed on a grounded ( $V = 0$  Volt) conducting plane with a tiny gap separating it from the plane. The cone is maintained at a voltage  $V_0$  Volts.

a) Use Laplace's equation to find the electric potential  $V(r, \theta, \phi)$  in the region  $\theta_0 < \theta < \frac{\pi}{2}$ . You might need the following in your calculation:

$$\int \frac{d\theta}{\sin \theta} = \ln(\tan \frac{\theta}{2}) + C, \text{ where } C \text{ is a constant, and}$$

$$\tan \frac{\pi}{4} = 1.$$

**Solution:**

Starting with Laplace's equation in spherical coordinates in the region of interest, and making use of the symmetry [ $\partial(\dots)/\partial r = \partial(\dots)/\partial \phi = 0$ ], we get

$$\nabla^2 V = \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} [\sin \theta \frac{\partial V}{\partial \theta}] = 0. \tag{3}$$

Integrating this once, we get

$$\frac{\partial V}{\partial \theta} = \frac{A}{\sin \theta}, \tag{4}$$

where  $A$  is a constant. Integrating one more time, we get

$$V = A \ln[\tan \frac{\theta}{2}] + B, \tag{5}$$

where  $B$  is another constant. Using the two boundary conditions -  $V(\theta = 0) = 0$  and  $V(\theta = \theta_0) = V_0$ , we find the two constants, and then the voltage is given by

$$V(r, \theta, \phi) = V(\theta) = V_0 \frac{\ln[\tan \frac{\theta}{2}]}{\ln[\tan \frac{\theta_0}{2}]} \quad (6)$$

b) Calculate the electric field  $\mathbf{E}$  in the region  $\theta_0 < \theta < \frac{\pi}{2}$  using your result from part (a).

**Solution:**

The electric field is given by  $\mathbf{E} = -\nabla V = -\frac{1}{r} \frac{\partial V}{\partial \theta} \mathbf{a}_\theta$ . This leads to

$$\mathbf{E} = -\frac{V_0}{r \sin \theta \ln[\tan \frac{\theta_0}{2}]} \mathbf{a}_\theta, \quad (7)$$

which indicates that the electric field points along the  $\mathbf{a}_\theta$  direction from the conical conductor to the planar conductor (remember the  $\ln$  factor is negative!). The electric field gets stronger as we approach the tip of the cone ( $r \rightarrow 0$ ), and weakens as we get farther from the top  $|E| \rightarrow 0$  as  $r \rightarrow \infty$ . Clearly this implies that there is a charge pileup near the tip of the cone.

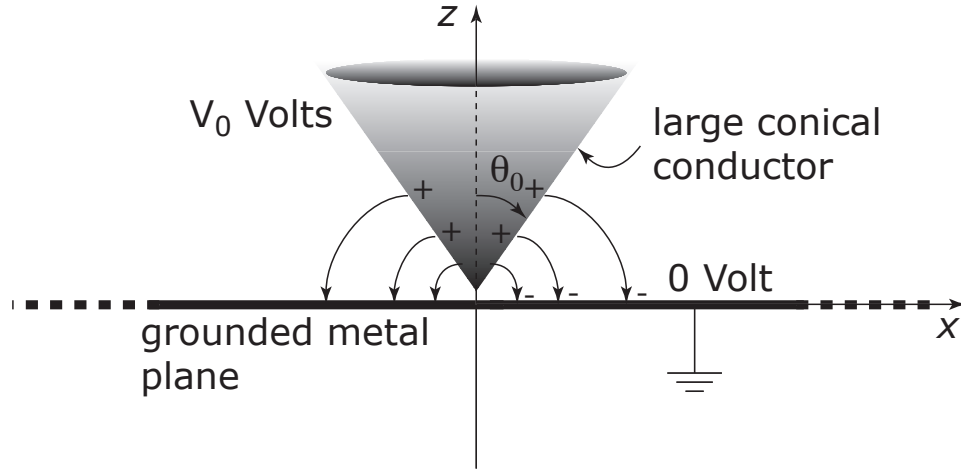


Figure 4: Problem (4)

c) Find the charge density on the two conductors from your result of part (b).

**Solution:**

The sheet charge density  $\rho_s$  on the metallic conductors is given by the boundary condition  $\mathbf{D} \cdot \mathbf{n} = \rho_s$ . For the conical conductor, the charge is positive since the field lines originate from there (assuming  $V_0 > 0$ ). It is given by

$$\rho_s^{cone}(r, \theta, \phi) = \epsilon_0 \mathbf{E}(r, \theta = \theta_0, \phi) \cdot \mathbf{a}_\theta = -\frac{\epsilon_0 V_0}{r \sin \theta_0 \ln[\tan \frac{\theta_0}{2}]}, \quad (8)$$

and the corresponding charge on the planar conductor is given by

$$\rho_s^{plane}(r, \theta, \phi) = \epsilon_0 \mathbf{E}(r, \theta = \frac{\pi}{2}, \phi) \cdot -\mathbf{a}_\theta = \frac{\epsilon_0 V_0}{r \ln[\tan \frac{\theta_0}{2}]}.$$
 (9)

The charge and the electric field lines are sketched in Figure 4.

**End.**

(Formulae in the next page...)

## Formulae

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### A) Fundamental constants

Permittivity of vacuum:  $\epsilon_0 \approx \frac{1}{36\pi} \times 10^{-9} \text{F/m}$ ,

Permeability of vacuum:  $\mu_0 = 4\pi \times 10^{-7} \text{H/m}$ .

Speed of light in free space:  $c = 1/\sqrt{\epsilon_0\mu_0} \approx 3 \times 10^8 \text{ m/s}$ .

Speed of light in non-conductive media:  $v = c/\sqrt{\epsilon_r\mu_r}$ .

### B) Maxwell's Equations: [Integral Form], [Differential Form]

Gauss's Law for Electric Fields :  $[\oint_s \mathbf{D} \cdot d\mathbf{S} = \int_v \rho_v dv = Q_{encl}]$  ,  $[\nabla \cdot \mathbf{D} = \rho_v]$

Gauss's Law for Magnetic Fields :  $[\oint_s \mathbf{B} \cdot d\mathbf{S} = 0]$  ,  $[\nabla \cdot \mathbf{B} = 0]$

Faraday's Law :  $[\oint_c \mathbf{E} \cdot d\mathbf{l} = -\frac{d}{dt}(\int_s \mathbf{B} \cdot d\mathbf{S})]$  ,  $[\nabla \times \mathbf{E} = -\frac{d\mathbf{B}}{dt}]$

Ampere's Law :  $[\oint_c \mathbf{H} \cdot d\mathbf{l} = \int_s \mathbf{J} \cdot d\mathbf{S} + \frac{d}{dt}(\int_s \mathbf{D} \cdot d\mathbf{S})]$  ,  $[\nabla \times \mathbf{H} = \mathbf{J} + \frac{d\mathbf{D}}{dt}]$

Charge continuity equation:  $\nabla \cdot \mathbf{J} + \frac{\partial \rho_v}{\partial t} = 0$ .

In all of the above,  $\mathbf{D} = \epsilon_0\epsilon_r\mathbf{E}$  and  $\mathbf{B} = \mu_0\mu_r\mathbf{H}$ , where  $\epsilon_r$  is the relative dielectric constant and  $\mu_r$  is the relative permeability of the material medium.

### C) Wave propagation characteristics in material medium

The conduction current density that adds to Ampere's law above is given by  $\mathbf{J}_c = \sigma\mathbf{E}$ , where  $\sigma$  is the conductivity of the material medium.

In addition to Maxwell's equations above in (B), the boundary conditions are -

For electric fields,  $\mathbf{n} \cdot (\mathbf{D}_1 - \mathbf{D}_2) = \rho_s$ , and  $\mathbf{n} \times (\mathbf{E}_1 - \mathbf{E}_2) = 0$ .

For magnetic fields,  $\mathbf{n} \cdot (\mathbf{B}_1 - \mathbf{B}_2) = 0$  and  $\mathbf{n} \times (\mathbf{H}_1 - \mathbf{H}_2) = \mathbf{J}_s$ .

In a generic material medium characterized by the parameters  $(\epsilon_r, \mu_r, \sigma)$ , the phasor notation of the electric field component of an EM wave moving in the  $+z$  direction can be written as  $\hat{E} = \hat{E}_m e^{-\hat{\gamma}z}$ , where

$$\hat{\gamma} = \alpha + j\beta,$$

$$\alpha = \omega \sqrt{\frac{\mu\epsilon}{2} [\sqrt{1 + (\frac{\sigma}{\omega\epsilon})^2} - 1]}^{1/2},$$

$$\beta = \omega \sqrt{\frac{\mu\epsilon}{2} [\sqrt{1 + (\frac{\sigma}{\omega\epsilon})^2} + 1]}^{1/2},$$

and the corresponding  $\hat{H}$  is related to the electric field component by the complex impedance  $\hat{H} = \hat{E}/\hat{\eta}$ , where

$$\hat{\eta} = \sqrt{\frac{\mu}{\epsilon - j\frac{\sigma}{\omega}}} = \frac{\sqrt{\frac{\mu}{\epsilon}}}{[1 + (\frac{\sigma}{\omega\epsilon})^2]^{1/4}} e^{j\frac{1}{2} \tan^{-1}(\frac{\sigma}{\omega\epsilon})}.$$

The skin depth of a conductive medium is  $\delta = 1/\alpha$ .

The total energy stored in the electric field is  $W_E = \int_v \frac{1}{2}\epsilon|\mathbf{E}|^2 dv$  and in the magnetic field is  $W_M = \int_v \frac{1}{2}\mu|\mathbf{H}|^2 dv$ , and therefore the total energy stored in a volume with both electric and magnetic fields is given by  $W = W_E + W_M$ .

Power is transported by an EM wave, and the Poynting vector is defined as  $\mathbf{P} = \mathbf{E} \times \mathbf{H}$ .

The time-averaged power transported by an EM wave propagating in the  $+z$  direction is given by  $\mathbf{P}_{av} = \frac{|\hat{E}_m|^2}{2\eta_0} \mathbf{a}_z$  in vacuum or air, and by  $\mathbf{P}_{av} = \frac{|\hat{E}_m|^2}{2\hat{\eta}} e^{-2\alpha z} \cos\theta \mathbf{a}_z$ , where  $\theta = \frac{1}{2} \tan^{-1}(\frac{\sigma}{\omega\epsilon})$  in a conductive medium.

## D) Static Electric and Magnetic Fields

• Under static conditions, the Electric field is a conservative field, and therefore can be defined as the gradient of a scalar electric potential, i.e.,

$$\mathbf{E} = -\nabla V$$

and equivalently, the potential difference between two points can be uniquely determined by the line integral of the electric field:

$$V_{ab} = \int_a^b \mathbf{E} \cdot d\mathbf{l}.$$

• The electric potential due to point, line, sheet, and volume charges are given by, respectively,  $V_{point} = \frac{Q}{4\pi\epsilon r}$ ,  $V_{line} = \int \frac{\rho_l dl}{4\pi\epsilon r}$ ,  $V_{sheet} = \int \frac{\rho_s ds}{4\pi\epsilon r}$ ,  $V_{vol} = \int \frac{\rho_v dv}{4\pi\epsilon r}$ , where  $l, s, v$  stand for line, sheet, and volume respectively.

• The electric potential follows the principle of superposition; for example, for  $N$  point charges  $Q_i$  each located at  $r_i$ , the total potential is

$$V = \sum_{i=1}^N \frac{Q_i}{4\pi\epsilon r_i}.$$

• The capacitance of an object is a geometrical property that is defined as the ratio of the positive charge to the resulting potential difference between the conductors, i.e.,

$$C = \frac{Q}{V}.$$

• Gauss's law for electric field may now be re-cast in terms of the electric potential; this leads to Poisson's and Laplace's equations:

$$\nabla^2 V = -\frac{\rho_v}{\epsilon} \text{ (Poisson's equation), \&}$$

$$\nabla^2 V = 0 \text{ (Laplace's equation, valid if } \rho_v = 0).$$

• A scalar field for the Magnetic field is not possible. However, the Magnetic field  $\mathbf{B}$  may be re-cast in terms of a Magnetic vector potential  $\mathbf{A}$ , such that

$$\mathbf{B} = \nabla \times \mathbf{A}, \text{ and}$$

$$\mathbf{A} = \frac{\mu}{4\pi} \int \frac{\mathbf{J}(\mathbf{r}') dv'}{R}, \text{ where } \mathbf{J}(\mathbf{r}') \text{ is the current density which produces the magnetic field.}$$

• Magnetic circuits may easily be solved by using the analogies with electrical circuits. The analogies are :

$$V \leftrightarrow NI \text{ (magnetomotive force),}$$

$$I \leftrightarrow \psi_m = B \cdot S, \text{ (} \psi_m \text{: magnetic flux, } B \text{: magnetic field, } S \text{: cross-sectional area),}$$

$$R \leftrightarrow R = \frac{L}{\mu S}, \text{ } L \text{: length of magnetic core, } \mu \text{: permeability,}$$

$$\text{and } \sigma \leftrightarrow \mu.$$

The analogies are valid as long as the permeability of the core is large enough to prevent substantial flux leakage out of the core.

• The ability of an object to produce magnetic flux in response to the current flowing through it is called the self inductance of the object, and is defined as

$$L_{11} = \frac{N_1 \psi_{11}}{I_1}.$$

Similarly, the mutual inductance between two conductors is given by

$$L_{12} = \frac{N_2 \psi_{12}}{I_1}.$$

## E) Line, Surface, and Volume vector differential elements

Rectangular: (Unit Vectors:  $\mathbf{a}_x, \mathbf{a}_y, \mathbf{a}_z$ )

$$d\mathbf{l} = dx\mathbf{a}_x + dy\mathbf{a}_y + dz\mathbf{a}_z,$$

$$ds_x = dydz\mathbf{a}_x, ds_y = dx dz\mathbf{a}_y, ds_z = dx dy\mathbf{a}_z,$$

$$dv = dx dy dz.$$

Cylindrical: (Unit Vectors:  $\mathbf{a}_\rho, \mathbf{a}_\phi, \mathbf{a}_z$ )

$$d\mathbf{l} = d\rho\mathbf{a}_\rho + \rho d\phi d\mathbf{y}\mathbf{a}_\phi + dz\mathbf{a}_z,$$

$$ds_\rho = \rho d\phi dz\mathbf{a}_\rho, ds_\phi = d\rho dz\mathbf{a}_\phi, ds_z = \rho d\rho d\phi\mathbf{a}_z,$$

$$dv = \rho d\rho d\phi dz.$$

Spherical: (Unit Vectors:  $\mathbf{a}_r, \mathbf{a}_\theta, \mathbf{a}_\phi$ )

$$d\mathbf{l} = dr\mathbf{a}_r + r d\theta\mathbf{a}_\theta + r \sin\theta d\phi\mathbf{a}_\phi,$$

$$ds_r = r^2 \sin\theta d\theta d\phi\mathbf{a}_r, ds_\theta = r \sin\theta dr d\phi\mathbf{a}_\theta, ds_\phi = r dr d\theta\mathbf{a}_\phi,$$

$$dv = r^2 \sin\theta dr d\theta d\phi.$$

## F) div, grad, & curl expressions in various coordinate systems

Use the metric coefficients from the Table below.

For a scalar field  $\Phi$ ,

$$\text{grad } \Phi = \nabla\Phi = \frac{1}{h_1} \frac{\partial\Phi}{\partial u_1} \mathbf{a}_1 + \frac{1}{h_2} \frac{\partial\Phi}{\partial u_2} \mathbf{a}_2 + \frac{1}{h_3} \frac{\partial\Phi}{\partial u_3} \mathbf{a}_3, \text{ and}$$

$$\text{Laplacian } \Phi = \nabla^2\Phi = \frac{1}{h_1 h_2 h_3} \left[ \frac{\partial}{\partial u_1} \left( \frac{h_2 h_3}{h_1} \frac{\partial\Phi}{\partial u_1} \right) + \frac{\partial}{\partial u_2} \left( \frac{h_1 h_3}{h_2} \frac{\partial\Phi}{\partial u_2} \right) + \frac{\partial}{\partial u_3} \left( \frac{h_1 h_2}{h_3} \frac{\partial\Phi}{\partial u_3} \right) \right].$$

For a vector field  $\mathbf{A} = A_1\mathbf{a}_1 + A_2\mathbf{a}_2 + A_3\mathbf{a}_3$ ,

$$\text{div } \mathbf{A} = \frac{1}{h_1 h_2 h_3} \left[ \frac{\partial(A_1 h_2 h_3)}{\partial u_1} + \frac{\partial(A_2 h_1 h_3)}{\partial u_2} + \frac{\partial(A_3 h_1 h_2)}{\partial u_3} \right], \text{ and}$$

$$\text{curl}(\mathbf{A}) = \begin{vmatrix} \frac{\mathbf{a}_1}{h_2 h_3} & \frac{\mathbf{a}_2}{h_1 h_3} & \frac{\mathbf{a}_3}{h_1 h_2} \\ \frac{\partial}{\partial u_1} & \frac{\partial}{\partial u_2} & \frac{\partial}{\partial u_3} \\ h_1 A_1 & h_2 A_2 & h_3 A_3 \end{vmatrix}.$$

	Cartesian	Cylindrical	Spherical
Independent Variables ( $u_1, u_2, u_3$ )	$x, y, z$	$\rho, \phi, z$	$r, \theta, \phi$
Vector components ( $A_1, A_2, A_3$ )	$A_x, A_y, A_z$	$A_\rho, A_\phi, A_z$	$A_r, A_\theta, A_\phi$
Unit Vectors ( $\mathbf{a}_1, \mathbf{a}_2, \mathbf{a}_3$ )	$\mathbf{a}_x, \mathbf{a}_y, \mathbf{a}_z$	$\mathbf{a}_\rho, \mathbf{a}_\phi, \mathbf{a}_z$	$\mathbf{a}_r, \mathbf{a}_\theta, \mathbf{a}_\phi$
Metric Coefficients ( $h_1, h_2, h_3$ )	1, 1, 1	1, $\rho$ , 1	1, $r$ , $r \sin\theta$