

# EE566 Solid State Devices

Spring 2005

Dept of Electrical Engineering

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## Assignment 3 SOLUTIONS

Assignment 3 - EE 566 SOLID STATE DEVICES, SPRING 2005 02/16/05

SOLUTIONS

PROBLEM 1 (3-7, MKC). See the charge-field-band diagram.

Electric field:  $E(0) = E_{max}$

$$\frac{1}{2} E_{max} x_n(V_a) = \phi_i - V_a$$

$$OACD \rightarrow \frac{1}{2} [E_{max} + E(t)] \cdot t = \phi_B$$

$$E(x) = E_{max} \left(1 - \frac{x}{x_n(V_a)}\right) \rightarrow E(t) = E_{max} \left(1 - \frac{t}{x_n(V_a)}\right)$$

$$\therefore E_{max} \left[2 - \frac{t}{x_n(V_a)}\right] = \frac{2\phi_B}{t}$$

total Area OAB =  $(\phi_i - V_a) = \frac{1}{2} E_{max} x_n(V_a)$

Area OACD =  $\phi_B$

Also,  $x_n(V_a) = \frac{\epsilon_s E_{max}}{q N_D} = \frac{2(\phi_i - V_a)}{E_{max}}$

Therefore, we get

$$E_{max} \left[2 - \frac{t \cdot E_{max}}{2(\phi_i - V_a)}\right] = \frac{2\phi_B}{t}$$

$$E_{max}^2 - \frac{4(\phi_i - V_a)}{t} E_{max} + \frac{4\phi_B(\phi_i - V_a)}{t^2} = 0$$

$$\therefore E_{max} = \frac{2(\phi_i - V_a)}{t} \left[1 \pm \sqrt{1 - \frac{\phi_B(\phi_i - V_a)}{t^2}}\right]$$

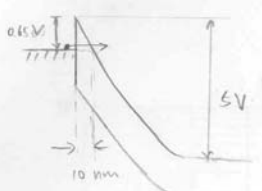
→ ONLY REASONABLE SOLN:  
 $E_{max} = 6.72 \times 10^5 \text{ V/cm}$   
 (Other is  $E_{max}' = 1.93 \times 10^7 \text{ V/cm} \gg E_{breakdown}$ )

$$\Rightarrow N_D = \frac{\epsilon_s E_{max}^2}{2q(\phi_i - V_a)} \sim 3 \times 10^{17} / \text{cm}^3$$

b) From fig 1-15 in MKC, for  $N_D \sim 3 \times 10^{17} / \text{cm}^3$ ,  $f \sim 5 \times 10^{-2} \text{ SL/cm}$

c) Show above.

PROBLEM 2 - Solution by A. Wang -



$$\frac{df}{dV} = -\frac{1}{4\pi\sqrt{LC}} \frac{1}{c} \frac{dc}{dV} = 0.22 \times 10^6 \text{ Hz/V}$$

$$\therefore N(x_d) = \frac{(c/A)^3}{q\epsilon_s d(c/A)/dV} = \frac{c^3}{A^2 N(x_d) q \epsilon_s} = \frac{dc}{dV}$$

$$\frac{-1}{4\pi\sqrt{LC}} \frac{1}{c} \frac{c^3}{A^2 N(x_d) q \epsilon_s} = -0.22 \times 10^6$$

$$\frac{c^2}{A^2 4\pi\sqrt{LC} q \epsilon_s \times 0.22 \times 10^6} = N(x_d)$$

$$c = A \cdot \epsilon_s / x_d$$

$$\frac{A^2 \epsilon_s^2 / x_d^2}{A^2 4\pi\sqrt{LC} q \epsilon_s \times 0.22 \times 10^6} = N(x_d)$$

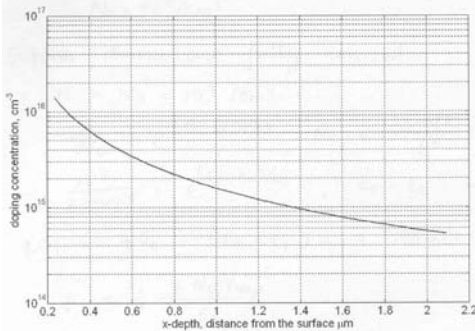
$$\frac{(1.1 \times 10^{-12})^2 / x_d^2}{4 \times 3.14 \times \sqrt{2 \times 10^{-3} \times 10^{-7}} / 1.6 \times 10^{-19} \times 0.22 \times 10^6} = N(x_d)$$

$$1.58 \times 10^{12} x_d^{-2} = N(x_d)$$

at  $V = 0$ ,  $c = 41.8 \times 10^{-12} = A \cdot \frac{\epsilon_s}{x_d}$ ,  $x_d = 2.33 \times 10^{-7} \text{ m} = 233 \text{ nm}$

$c = 4.65 \times 10^{-12} = A \cdot \frac{\epsilon_s}{x_d}$ ,  $x_d = 2.09 \times 10^{-6} \text{ m} = 2.09 \mu\text{m}$

The doping profile is plotted on the next page.



### PROBLEM 3 - Solution by Jengxiao Jin -

Problem3

a) We know that the thickness of depletion area can be calculated using equation

$$\frac{q^2 N_D x_d^2}{2\epsilon_s} + kT \ln\left(\frac{N_c}{N_D}\right) = q\phi_B$$

Now,  $q\phi_B = 1\text{eV}$ ,  $N_D = 1\text{e}17/\text{cm}^3$ ,  $N_c = 2.2\text{e}19/\text{cm}^3$ , so we can get the depletion distance

$$x_d = 1.21\text{e-}7\text{ m}$$

For such a depletion area, we know that the highest electric field exists at the junction, and its value is

$$E = \frac{qN_D x_d}{\epsilon_{\text{GaN}}} = \frac{1.6 \times 10^{-19} \times 10^{23} \times 1.21 \times 10^{-7}}{8.9 \times 8.854 \times 10^{-12}} = 2.5 \times 10^7 \text{ V/m}$$

The breakdown field for GaN is about  $5\text{e}6\text{ V/cm}$ , that is  $5\text{e}8\text{V/m}$ , so the electric field at the surface of the semiconductor is smaller than the breakdown voltage  $F_{\text{BR}}$ .

b) Now we know that as reverse voltage changes, the capacitance doesn't change, which means the depletion area also doesn't change, so we can know that before reverse bias voltage is added, the whole GaN area has already been depleted, thus we know that

$$t_{\text{cap}} < x_{\text{depl}} = 1.21 \times 10^{-7} \text{ m}$$

If we know the value of the capacitance, then we can get the value of  $t_{\text{cap}}$  by doing calculation

$$t_{\text{cap}} = \frac{C}{\epsilon_{\text{GaN}}}$$

c) To achieve breakdown, we should have an electric field that equals to  $F_{\text{BR}}$  at the surface of the semiconductor. Under reverse bias  $V_R$ , we can calculate the electric field

$$E = \frac{qN_D t_{\text{cap}}}{\epsilon_{\text{GaN}}} + \frac{V_R}{t_{\text{cap}}}$$

Let  $E$  equals breakdown field value  $F_{\text{BR}} = 5\text{e}6\text{ V/cm}$  when  $V_R = 38.4\text{V}$ , and we can get the value of  $t_{\text{cap}}$ .

$$t_{\text{cap}} = 8 \times 10^{-8} \text{ m} \text{ or } 2.42 \times 10^{-6} \text{ m}$$

Since  $t_{\text{cap}} < x_{\text{depl}} = 1.21\text{e-}7\text{m}$ , we can get the value of  $t_{\text{cap}}$  is  $8\text{e-}8\text{ m}$  or  $80\text{nm}$ .

The Charge-Field\_Band diagrams are simple and should be drawn in a complete solution. Remember "Kroemer's Lemma of Proven Ignorance"...

PROBLEM 4 - Comments by me and a possible solution by L. Chuanxin -

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PROBLEM 4

As you must have realized, the desired specifications are rather difficult to satisfy. Requirement (b) for the Schottky diode to be unaffected by visible light rules out Si, Ge, and almost all Zinc-Blende III-V semiconductors, since their bandgaps lie in or below the visible energies ( $\sim 2-3\text{eV}$  or  $400\text{nm}-800\text{nm}$ ).

Therefore, one has to opt for a wide-bandgap semiconductor - GaN,  $\text{Al}_x\text{Ga}_{1-x}\text{N}$ , AlN, or ZnO are some possible choices. Diamond is another attractive choice. Typically, wide bandgap semiconductors are difficult to dope p-type.

If one is able to dope GaN p-type, it might be possible to get a surface barrier height of  $\sim 2\text{Volts}$  with certain metals. However, in real world, there are certain constraints. (See the attached solutions...)

For an ideal GaN - Metal contact, (No surface pinning),  
(if  $\phi_M = 6.35\text{eV}$ )  
n-GaN - Pt can give the required condition, but this is rarely seen in practice...

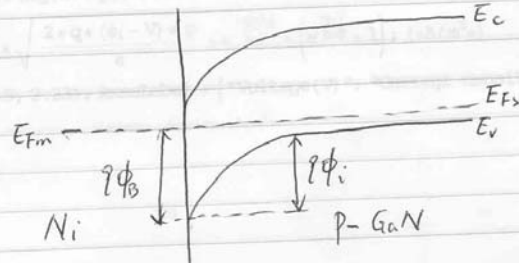
Another option suggested was to design the Schottky diode such that appreciable 'tunneling' occurs @  $2\text{Volts}$ . However, tunneling in a Schottky occurs mainly to carry reverse-bias current...

There might be other possible ways too...

Problem 4.

metal : Ni  
 Semiconductor : P-GaN  
 Acceptor density :  $\sim 10^{20} \text{ cm}^{-3}$       Hole concentration :  $\sim 10^{18} \text{ cm}^{-3}$   
 Acceptor : Mg

$$\chi_{\text{Ni}} = 5.15 \text{ eV} \quad \chi_{\text{GaN}} = 4.1 \text{ eV}$$



$$\phi_B = \chi + E_g - \phi_i = 2.39 \text{ eV}$$

The acceptors with the doping concentration of around  $10^{20} \text{ cm}^{-3}$  can generate holes which have the concentration of approximately  $10^{18} \text{ cm}^{-3}$

$$\phi_i = \phi_B - \frac{kT}{q} \ln\left(\frac{N_v}{P}\right) \approx 2.3 \text{ V}$$

- $\therefore$  a)  $V_{\text{turn-on}} \approx \phi_i \approx 2 \text{ V}$   
 b)  $E_g(\text{GaN}) = 3.4 \text{ eV} > \text{energy of visible light}$

(\*Ni/p-GaN\*)

$$q = 1.6 \times 10^{-19}; (*C*)$$

$$d = 4 \times 10^{-4}; (*m^2/S*)$$

$$nv = 4.6 \times 10^{19} \times 10^6; (*m^3*)$$

$$k = 1.38 \times 10^{-23}; (*J/K*)$$

$$T = 300; (*K*)$$

$$p = 10^{18} \times 10^6; (*m^3*)$$

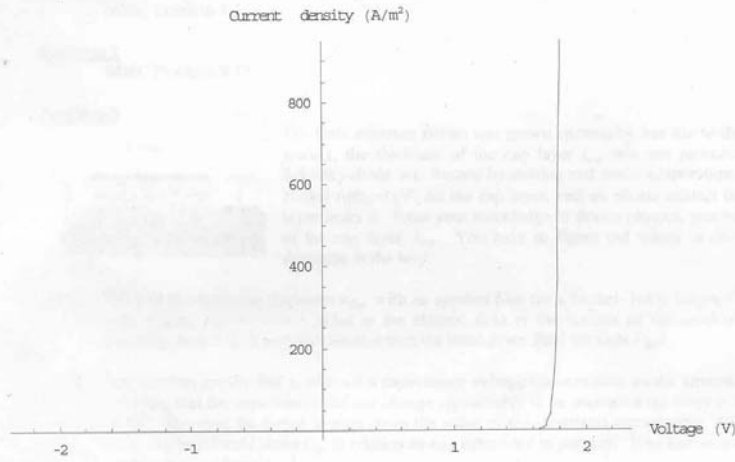
$$\epsilon = 9 \times 8.85 \times 10^{-12}; (*F/m*)$$

$$\phi_B = 4.1 + 3.44 - 5.15; (*V*)$$

$$\phi_i = \phi_B - 0.026 \times \text{Log}[nv / p]; (*V*)$$

$$J = \frac{q^2 \cdot d \cdot nv}{k \cdot T} \cdot \sqrt{\frac{2 \cdot q \cdot (\phi_i - V) \cdot p}{e}} \cdot e^{-\frac{q \cdot \phi_B}{k \cdot T}} \cdot \left( e^{\frac{q \cdot V}{k \cdot T}} - 1 \right); (*A/m^2*)$$

Plot[J, {V, -5, 2.23}, AxesLabel -> {"Voltage(V)", "Current density(A/m^2)"}]



$$E_{BR} = 5 \times 10^6 \text{ V/cm for GaN}$$

$$\chi_d = \sqrt{\frac{2 \epsilon_s \phi_i}{q \cdot p}} = \sqrt{\frac{2 \times 9 \times 8.85 \times 10^{-14} \times 2.3}{1.6 \times 10^{-19} \times 10^{18}}} = 47.9 \text{ nm}$$

$$|V_{BR}| = \frac{1}{2} E_{BR} \chi_d = \frac{1}{2} \times 5 \times 10^6 \times 47.9 \times 10^{-7} \approx 12 \text{ V}$$

$$\boxed{V_{BR} \approx 12 \text{ V}}$$